# SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer 

General Description

The MAX2042 single, high-linearity upconversion/downconversion mixer provides +36 dBm IIP3, 7.3 dB noise figure, and 7.2 dB conversion loss for 2000 MHz to 3000 MHz WCS, LTE, WiMAXTM, and MMDS wireless infrastructure applications. With a wide LO frequency range of 1800 MHz to 2800 MHz , this particular mixer is ideal for low-side LO injection receiver and transmitter architectures. High-side LO injection is supported by the MAX2042A, which is pinpin and functionally compatible with the MAX2042.
In addition to offering excellent linearity and noise performance, the MAX2042 also yields a high level of component integration. This device includes a doublebalanced passive mixer core, an LO buffer, and on-chip baluns that allow for single-ended RF and LO inputs. The MAX2042 requires a nominal LO drive of 0 dBm , and supply current is typically 138 mA at $\mathrm{V}_{\mathrm{CC}}=+5.0 \mathrm{~V}$ or 120 mA at $\mathrm{VCC}=+3.3 \mathrm{~V}$.
The MAX2042 is pin compatible with the MAX2042A 2000MHz to 3900 MHz mixer. The device is also pin similar with the MAX2029/MAX2031 650MHz to 1000 MHz mixers, the MAX2039/MAX2041 1700MHz to 3000 MHz mixers, and the MAX2044/MAX2044A 3000 MHz to 4000 MHz mixers, making this entire family of up/downconverters ideal for applications where a common PCB layout is used for multiple frequency bands.
The MAX2042 is available in a compact 20-pin thin QFN ( $5 \mathrm{~mm} \times 5 \mathrm{~mm}$ ) package with an exposed pad. Electrical performance is guaranteed over the extended $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ temperature range.

## Applications

2.3GHz WCS Base Stations
2.5GHz WiMAX and LTE Base Stations
2.7GHz MMDS Base Stations

Fixed Broadband Wireless Access
Wireless Local Loop
Private Mobile Radios
Military Systems
Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE |
| :--- | :--- | :--- |
| MAX2042ETP + | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 20 Thin QFN-EP* |
| MAX2042ETP +T | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 20 Thin QFN-EP* |

+ Denotes a lead(Pb)-free/RoHS-compliant package.
*EP = Exposed pad.
$T$ = Tape and reel.

Features

- 2000 MHz to 3000 MHz RF Frequency Range
- 1800 MHz to 2800 MHz LO Frequency Range
- 50 MHz to 500 MHz IF Frequency Range
- 7.2dB Conversion Loss
- 7.3dB Noise Figure
- +36dBm Typical IIP3
- +23.4dBm Typical Input 1dB Compression Point
- 70dBc Typical 2RF-2LO Spurious Rejection at PRF $=-10 \mathrm{dBm}$
- Integrated LO Buffer
- Integrated RF and LO Baluns for Single-Ended Inputs
- Low -3dBm to +3dBm LO Drive
- Pin Compatible with the MAX2042A 2000MHz to 3900MHz High-Side LO Injection Mixer
- Pin Similar with the MAX2029/MAX2031 650MHz to 1000MHz Mixers, MAX2039/MAX2041 1700MHz to 3000MHz Mixers, and MAX2044/MAX2044A 3000MHz to 4000 MHz Mixers
- Single +5.0 V or +3.3 V Supply
- External Current-Setting Resistor Provides Option for Operating Device in Reduced-Power/ReducedPerformance Mode

Pin Configuration/ Functional Diagram

## TOP VIEW


*EXPOSED PAD
WiMAX is a trademark of WiMAX Forum.

## SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer

## ABSOLUTE MAXIMUM RATINGS

| VCC to GND................................................................... 0.3 V to +5.5 VIF+, IF-, LOBIAS to GND................... -0.3 V to (VCC +0.3 V ) |  |
| :---: | :---: |
|  |  |
| RF, LO Input Power | +20dBm |
| RF, LO Current (RF and LO are DC shorted to GND through a balun). | 50 mA |
| Continuous Power Dissipation (Note 1) | 5.0W |
| OJA (Notes 2, 3) | $+38^{\circ} \mathrm{C} / \mathrm{W}$ |

$\operatorname{\theta JC}$ (Notes 1, 3).......................................................... $+13^{\circ} \mathrm{C} / \mathrm{W}$
Operating Case Temperature Range
(Note 4)......................................................... $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Junction Temperature .................................................... $150^{\circ} \mathrm{C}$
Storage Temperature Range.............................. $65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Lead Temperature (soldering, 10s) ................................ $+300^{\circ} \mathrm{C}$

Note 1: Based on junction temperature $T_{J}=T_{C}+\left(\theta_{J C} \times V_{C C} \times I C C\right)$. This formula can be used when the temperature of the exposed pad is known while the device is soldered down to a PCB. See the Applications Information section for details. The junction temperature must not exceed $+150^{\circ} \mathrm{C}$.
Note 2: Junction temperature $T_{J}=T_{A}+\left(\theta_{J A} \times V_{C C} \times I C C\right)$. This formula can be used when the ambient temperature of the PCB is known. The junction temperature must not exceed $+150^{\circ} \mathrm{C}$.
Note 3: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.
Note 4: TC is the temperature on the exposed pad of the package. $\mathrm{T}_{\mathrm{A}}$ is the ambient temperature of the device and PCB.
Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## +5.0V SUPPLY DC ELECTRICAL CHARACTERISTICS

(Typical Application Circuit, $\mathrm{VCC}=+4.75 \mathrm{~V}$ to +5.25 V , no input AC signals. $\mathrm{TC}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$, unless otherwise noted. Typical values are at $\mathrm{VCC}=+5.0 \mathrm{~V}, \mathrm{TC}=+25^{\circ} \mathrm{C}$, all parameters are production tested.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply VoItage | VCC |  | 4.75 | 5.0 | 5.25 | V |
| Supply Current | ICC |  |  | 138 | 150 | mA |

## +3.3V SUPPLY DC ELECTRICAL CHARACTERISTICS

(Typical Application Circuit, $\mathrm{V}_{C C}=+3.0 \mathrm{~V}$ to +3.6 V , no input AC signals. $\mathrm{T}_{\mathrm{C}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$, unless otherwise noted. Typical values are at $\mathrm{V} \mathrm{CC}=+3.3 \mathrm{~V}, \mathrm{TC}=+25^{\circ} \mathrm{C}$, all parameters are production tested.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply VoItage | VCC |  | 3.0 | 3.3 | 3.6 | V |
| Supply Current | ICC |  |  | 120 | 135 | mA |

## RECOMMENDED AC OPERATING CONDITIONS

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RF Frequency Range |  | Typical Application Circuit with C1 $=8.2 \mathrm{pF}$, see Table 1 for details (Notes 5, 6) | 2000 |  | 3000 | MHz |
| LO Frequency | $f L O$ | (Notes 5, 6) | 1800 |  | 2800 | MHz |
| IF Frequency | fiF | Using M/A-Com MABAES0029 1:1 <br> transformer as defined in the Typical Application Circuit, IF matching components affect the IF frequency range (Notes 5, 6) | 50 |  | 500 | MHz |
| LO Drive | PLO | (Notes 5, 6) | -3 | 0 | +3 | dBm |

# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

## +5.0V SUPPLY AC ELECTRICAL CHARACTERISTICS (DOWNCONVERTER OPERATION)

(Typical Application Circuit with tuning elements outlined in Table 1, $\mathrm{VCC}=+4.75 \mathrm{~V}$ to +5.25 V , RF and LO ports are driven from $50 \Omega$ sources, PLO $=-3 \mathrm{dBm}$ to $+3 \mathrm{dBm}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{fRF}=2300 \mathrm{MHz}$ to $2900 \mathrm{MHz}, \mathrm{fIF}=300 \mathrm{MHz}, \mathrm{fLO}=2000 \mathrm{MHz}$ to 2600 MHz , fRF $>\mathrm{fLO}$, $\mathrm{T}_{\mathrm{C}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are for $\mathrm{T} \mathrm{C}=+25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+5.0 \mathrm{~V}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}, \mathrm{fRF}=2300 \mathrm{MHz}, \mathrm{fLO}=2300 \mathrm{MHz}$, $\mathrm{fIF}^{2}=300 \mathrm{MHz}$. All parameters are guaranteed by design and characterization, unless otherwise noted.) (Note 7)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Small-Signal Conversion Loss | LC | fRF $=2300 \mathrm{MHz}$ to $2900 \mathrm{MHz}, \mathrm{TC}=+25^{\circ} \mathrm{C}$ (Note 8) |  | 6.7 | 7.2 | 8.1 | dB |
| Loss Variation vs. Frequency | $\Delta \mathrm{LC}$ | fRF $=2305 \mathrm{MHz}$ to 2360 MHz |  |  | 0.15 |  |  |
|  |  | fRF $=2500 \mathrm{MHz}$ to 2570 MHz |  |  | 0.15 |  |  |
|  |  | fRF $=2570 \mathrm{MHz}$ to 2620MHz |  |  | 0.15 |  | dB |
|  |  | fRF $=2500 \mathrm{MHz}$ to 2690 MHz |  |  | 0.15 |  |  |
|  |  | fRF $=2700 \mathrm{MHz}$ to 2900MHz |  |  | 0.20 |  |  |
| Conversion Loss Temperature Coefficient | TCCL | TC $=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |  |  | 0.0071 |  | $\mathrm{dB} /{ }^{\circ} \mathrm{C}$ |
| Single Sideband Noise Figure | NFSSB | No blockers present |  |  | 7.3 |  | dB |
| Noise Figure Temperature Coefficient | TCNF | fRF $=2300 \mathrm{MHz}$ to 2900 MHz , single sideband, no blockers present,$\mathrm{TC}=-40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C}$ |  |  | 0.019 |  | $\mathrm{dB} /{ }^{\circ} \mathrm{C}$ |
| Noise Figure Under Blocking | NFB | +8 dBm blocker to fRF $=2600 \mathrm{MHz}$, fL fBLOCKER = 2795 $\mathrm{VCC}=5.0 \mathrm{~V}, \mathrm{TC}=$ | e applied to RF port, $\begin{aligned} & \mathrm{O}=2300 \mathrm{MHz}, \\ & 1 \mathrm{~Hz}, \mathrm{PLO}=0 \mathrm{dBm}, \\ & \left.+25^{\circ} \mathrm{C} \text { (Notes 5, } 9\right) \end{aligned}$ |  | 20.8 | 25 | dB |
| Input 1dB Compression Point | IP1dB | $\begin{aligned} & \mathrm{TC}=+25^{\circ} \mathrm{C} \\ & (\text { Notes } 5,10) \end{aligned}$ | fRF $=2300 \mathrm{MHz}$ | 22.5 | 23.4 |  | dBm |
|  |  |  | fRF $=2600 \mathrm{MHz}$ | 20.6 | 22.1 |  |  |
|  |  |  | $f \mathrm{fF}=2900 \mathrm{MHz}$ | 17.6 | 20.7 |  |  |
| Third-Order Input Intercept Point | IIP3 | PRF1 $=$ PRF2 $=$ OdBm/tone,PLO = OdBm,$\mathrm{TC}=+25^{\circ} \mathrm{C}$ | $\begin{aligned} & \text { fRF1 }=2300 \mathrm{MHz}, \\ & \text { fRF2 }=2301 \mathrm{MHz}, \\ & \text { fLO }=2000 \mathrm{MHz}(\text { Note } 5) \end{aligned}$ | 34 | 36 |  | dBm |
|  |  |  | $\begin{aligned} & \text { fRF1 }=2600 \mathrm{MHz}, \\ & \text { fRF2 }=2601 \mathrm{MHz}, \\ & \text { fLO }=2300 \mathrm{MHz}(\text { Note } 8) \end{aligned}$ | 31 | 34 |  |  |
|  |  |  | $\begin{aligned} & \text { fRF1 }=2900 \mathrm{MHz}, \\ & \text { fRF2 }=2901 \mathrm{MHz}, \\ & \text { fLO }=2600 \mathrm{MHz} \text { (Note 5) } \end{aligned}$ | 28 | 30 |  |  |
| IIP3 Variation with TC |  | $\begin{aligned} & \text { fRF }=2300 \mathrm{MHz} \text { to } \\ & \text { fRF1 }- \text { fRF2 }=1 \mathrm{MH} \\ & \text { tone, } \mathrm{TC}=-40^{\circ} \mathrm{C} \end{aligned}$ | $\begin{aligned} & 2900 \mathrm{MHz}, \\ & \text {, PRF1 = PRF2 }=0 \mathrm{dBm} / \\ & 0+85^{\circ} \mathrm{C} \end{aligned}$ |  | $\pm 0.5$ |  | dB |
| 2RF-2LO Spur Rejection | $2 \times 2$ | $\begin{aligned} & \text { fSPUR = fLO + } \\ & 150 \mathrm{MHz} \text { (Note 5) } \end{aligned}$ | PRF $=-10 \mathrm{dBm}$ | 64 | 70 |  | dBc |
|  |  |  | $\mathrm{PRF}=0 \mathrm{dBm}$ | 54 | 60 |  |  |
| 3RF - 3LO Spur Rejection | $3 \times 3$ | $\begin{aligned} & \text { fSPUR }=\text { fLO }+ \\ & \text { 100MHz (Note 5) } \end{aligned}$ | PRF $=-10 \mathrm{dBm}$ | 80 | 92 |  | dBc |
|  |  |  | PRF $=0 \mathrm{dBm}$ | 60 | 72 |  |  |
| RF Input Return Loss | RLRF | LO on and IF terminated into a matched impedance |  |  | 17 |  | dB |
| LO Input Return Loss | RLLO | RF and IF terminated into a matched impedance |  |  | 15 |  | dB |

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## +5.0V SUPPLY AC ELECTRICAL CHARACTERISTICS (DOWNCONVERTER OPERATION) (continued)

(Typical Application Circuit with tuning elements outlined in Table 1, $\mathrm{V}_{\mathrm{CC}}=+4.75 \mathrm{~V}$ to +5.25 V , RF and LO ports are driven from $50 \Omega$ sources, PLO $=-3 \mathrm{dBm}$ to +3 dBm , PRF $=0 \mathrm{dBm}, \mathrm{fRF}=2300 \mathrm{MHz}$ to $2900 \mathrm{MHz}, \mathrm{fIF}=300 \mathrm{MHz}, \mathrm{fLO}=2000 \mathrm{MHz}$ to 2600 MHz , fRF $>\mathrm{fLO}$, $\mathrm{TC}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are for $\mathrm{TC}=+25^{\circ} \mathrm{C}, \mathrm{VCC}=+5.0 \mathrm{~V}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}, \mathrm{fRF}=2300 \mathrm{MHz}, \mathrm{fLO}=2300 \mathrm{MHz}$, $\mathrm{fIF}_{\mathrm{IF}}=300 \mathrm{MHz}$. All parameters are guaranteed by design and characterization, unless otherwise noted.) (Note 7)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IF Output Impedance | ZIF | Nominal differential impedance at the IC's IF outputs |  | 50 |  | $\Omega$ |
| IF Output Return Loss | RLIF | RF terminated into $50 \Omega$, LO driven by $50 \Omega$ source, IF transformed to $50 \Omega$ using external components shown in the Typical Application Circuit |  | 18 |  | dB |
| RF-to-IF Isolation |  | PLO $=+3 \mathrm{dBm}$ (Note 8) | 30 | 37 |  | dB |
| LO Leakage at RF Port |  | $\mathrm{fLO}=2000 \mathrm{MHz}$ to 2800 MHz , PLO $=+3 \mathrm{dBm}$ (Note 8) |  | -28 | -22 | dBm |
| 2LO Leakage at RF Port |  | PLO $=+3 \mathrm{dBm}$ |  | -36 |  | dBm |
| LO Leakage at IF Port |  | $f L O=2000 \mathrm{MHz}$ to 2800 MHz , PLO $=+3 \mathrm{dBm}$ (Note 8) |  | -24.2 | -16 | dBm |

## +3.3V SUPPLY AC ELECTRICAL CHARACTERISTICS (DOWNCONVERTER OPERATION)

(Typical Application Circuit with tuning elements outlined in Table 1, RF and LO ports are driven from $50 \Omega$ sources. Typical values are for $\mathrm{TC}=+25^{\circ} \mathrm{C}, \mathrm{VCC}=+3.3 \mathrm{~V}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}, \mathrm{fRF}=2600 \mathrm{MHz}, \mathrm{fLO}=2300 \mathrm{MHz}, \mathrm{fIF}=300 \mathrm{MHz}$, unless otherwise noted.) (Note 7)

| PARAMETER | SYMBOL | CONDITIONS | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Small-Signal Conversion Loss | LC | (Note 8) | 7.2 |  | dB |
| Loss Variation vs. Frequency | $\Delta \mathrm{LC}$ | fRF $=2300 \mathrm{MHz}$ to 2900 MHz , any 100 MHz band | 0.2 |  | dB |
| Conversion Loss Temperature Coefficient | TCCL | TC $=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 0.008 |  | $\mathrm{dB} /{ }^{\circ} \mathrm{C}$ |
| Single Sideband Noise Figure | NFSSB | No blockers present | 7.5 |  | dB |
| Noise Figure Temperature Coefficient | TCNF | Single sideband, no blockers present, $\mathrm{TC}=-40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C}$ | 0.019 |  | $\mathrm{dB} /{ }^{\circ} \mathrm{C}$ |
| Input 1dB Compression Point | IP1dB | (Note 10) | 20 |  | dBm |
| Third-Order Input Intercept Point | IIP3 | $\begin{aligned} & \text { fRF1 }=2600 \mathrm{MHz}, \text { fRF2 }=2601 \mathrm{MHz}, \\ & \text { PRF1 }=\text { PRF2 }=0 \mathrm{dBm} / \text { tone } \end{aligned}$ | 31 |  | dBm |
| IIP3 Variation with TC |  | $\begin{aligned} & \text { fRF1 }=2600 \mathrm{MHz}, \text { fRF2 }=2601 \mathrm{MHz}, \\ & \text { PRF1 }=\text { PRF2 }=0 \mathrm{dBm} / \text { tone, } \\ & \text { TC }=-40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C} \end{aligned}$ | $\pm 0.25$ |  | dB |
| 2RF - 2LO Spur Rejection | $2 \times 2$ | PRF $=-10 \mathrm{dBm}, \mathrm{fSPUR}=\mathrm{fLO}+150 \mathrm{MHz}$ | 72 |  | dBc |
|  |  | PRF $=0 \mathrm{dBm}, \mathrm{fSPUR}=\mathrm{fLO}+150 \mathrm{MHz}$ | 62 |  |  |
| 3RF - 3LO Spur Rejection | $3 \times 3$ | PRF $=-10 \mathrm{dBm}, \mathrm{fSPUR}=f L \mathrm{O}+100 \mathrm{MHz}$ | 87 |  | dBc |
|  |  | PRF $=0 \mathrm{dBm}, \mathrm{fSPUR}=\mathrm{fLO}+100 \mathrm{MHz}$ | 67 |  |  |

# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

## +3.3V SUPPLY AC ELECTRICAL CHARACTERISTICS (DOWNCONVERTER OPERATION) (continued)

(Typical Application Circuit with tuning elements outlined in Table 1, RF and LO ports are driven from $50 \Omega$ sources. Typical values are for $\mathrm{T}_{\mathrm{C}}=+25^{\circ} \mathrm{C}, \mathrm{VCC}=+3.3 \mathrm{~V}$, $\mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, $\mathrm{fRF}=2600 \mathrm{MHz}, \mathrm{fLO}_{\mathrm{LO}}=2300 \mathrm{MHz}, \mathrm{f}_{\mathrm{IF}}=300 \mathrm{MHz}$, unless otherwise noted.) (Note 7)

| PARAMETER | SYMBOL | CONDITIONS | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RF Input Return Loss | RLRF | LO on and IF terminated into a matched impedance | 15 |  | dB |
| LO Input Return Loss | RLLO | RF and IF terminated into a matched impedance | 12 |  | dB |
| IF Output Impedance | ZIF | Nominal differential impedance at the IC's IF outputs | 50 |  | $\Omega$ |
| IF Output Return Loss | RLIF | RF terminated into $50 \Omega$, LO driven by $50 \Omega$ source, IF transformed to $50 \Omega$ using external components shown in the Typical Application Circuit | 18 |  | dB |
| Minimum RF-to-IF Isolation |  | fRF $=2300 \mathrm{MHz}$ to 2900 MHz , PLO $=+3 \mathrm{dBm}$ | 36 |  | dB |
| Maximum LO Leakage at RF Port |  | $\mathrm{fLO}=1800 \mathrm{MHz}$ to $2800 \mathrm{MHz}, \mathrm{PLO}=+3 \mathrm{dBm}$ | -24.5 |  | dBm |
| Maximum 2LO Leakage at RF Port |  | $\mathrm{fLO}=1800 \mathrm{MHz}$ to 2800 MHz , PLO $=+3 \mathrm{dBm}$ | -24 |  | dBm |
| Maximum LO Leakage at IF Port |  | $\mathrm{fLO}=1800 \mathrm{MHz}$ to $2800 \mathrm{MHz}, \mathrm{PLO}=+3 \mathrm{dBm}$ | -20 |  | dBm |

## +5.0V SUPPLY AC ELECTRICAL CHARACTERISTICS (UPCONVERTER OPERATION)

(Typical Application Circuit with tuning elements outlined in Table 2, $\mathrm{V}_{\mathrm{CC}}=+4.75 \mathrm{~V}$ to +5.25 V , RF and LO ports are driven from $50 \Omega$ sources, PLO $=-3 \mathrm{dBm}$ to $+3 \mathrm{dBm}, \mathrm{P}_{\mathrm{IF}}=0 \mathrm{dBm}, \mathrm{f}_{\mathrm{RF}}=2300 \mathrm{MHz}$ to $2900 \mathrm{MHz}, \mathrm{f}_{\mathrm{IF}}=200 \mathrm{MHz}, \mathrm{f}_{\mathrm{LO}}=2100 \mathrm{MHz}$ to 2700 MHz , fRF $>\mathrm{f}_{\mathrm{LO}}$, $\mathrm{TC}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$. Typical values are for $\mathrm{TC}=+25^{\circ} \mathrm{C}, \mathrm{VCC}=+5.0 \mathrm{~V}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}, \mathrm{fRF}=2600 \mathrm{MHz}, \mathrm{fLO}=2400 \mathrm{MHz}$, $\mathrm{f}_{\mathrm{IF}}=200 \mathrm{MHz}$. All parameters are guaranteed by design and characterization, unless otherwise noted.) (Note 7)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Small-Signal Conversion Loss | LC | (Note 8) |  | 6.8 |  | dB |
| Loss Variation vs. Frequency | $\Delta \mathrm{LC}$ | fRF $=2300 \mathrm{MHz}$ to 2960 MHz , any 100 MHz band |  | 0.2 |  | dB |
| Conversion Loss Temperature Coefficient | TCCL | TC $=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |  | 0.007 |  | $\mathrm{dB} /{ }^{\circ} \mathrm{C}$ |
| Input 1dB Compression Point | IP1dB | (Note 10) |  | 22.7 |  | dBm |
| Third-Order Input Intercept Point | IIP3 | $\begin{aligned} & \mathrm{fIF1}=200 \mathrm{MHz}, \mathrm{fIF} 2=201 \mathrm{MHz}, \\ & \text { PIF1 }=\text { PIF2 }=0 \mathrm{dBm} / \text { tone, fLO }=2400 \mathrm{MHz}, \\ & \text { PLO }=0 \mathrm{dBm}, \mathrm{TC}=+25^{\circ} \mathrm{C}(\text { Note } 8) \end{aligned}$ | 30 | 32.4 |  | dBm |
| IIP3 Variation with TC |  | $\begin{aligned} & \text { fIF1 }=200 \mathrm{MHz}, \mathrm{fIF} 2=201 \mathrm{MHz}, \\ & \text { PIF1 }=\text { PIF2 }=0 \mathrm{dBm} / \text { tone, fLO }=2400 \mathrm{MHz}, \\ & \text { PLO }=0 \mathrm{dBm}, \mathrm{TC}=-40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C} \end{aligned}$ |  | $\pm 0.5$ |  | dB |
| LO $\pm 2 \mathrm{IF}$ Spur Rejection | $1 \times 2$ | LO-2IF |  | 70 |  | dBc |
|  |  | LO + 2IF |  | 67 |  |  |
| $\mathrm{LO} \pm 3 \mathrm{IF}$ Spur Rejection | $1 \times 3$ | LO-3IF |  | 82 |  | dBc |
|  |  | LO + 3IF |  | 77 |  |  |
| Output Noise Floor |  | POUT $=0 \mathrm{dBm}$ ( Note 9) |  | -163 |  | dBm/Hz |

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## +3.3V SUPPLY AC ELECTRICAL CHARACTERISTICS (UPCONVERTER OPERATION)

(Typical Application Circuit with tuning elements outlined in Table 2, RF and LO ports are driven from $50 \Omega$ sources. Typical values are for $\mathrm{TC}=+25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=+3.3 \mathrm{~V}, \mathrm{P} \mathrm{PF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}, \mathrm{f}_{\mathrm{RF}}=2600 \mathrm{MHz}, \mathrm{f}_{\mathrm{LO}}=2400 \mathrm{MHz}, \mathrm{f}_{\mathrm{IF}}=200 \mathrm{MHz}$, unless otherwise noted.) (Note 7)

| PARAMETER | SYMBOL | CONDITIONS | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Small-Signal Conversion Loss | LC |  | 6.8 |  | dB |
| Loss Variation vs. Frequency | $\Delta \mathrm{LC}$ | fRF $=2300 \mathrm{MHz}$ to 2900 MHz , any 100 MHz band | 0.15 |  | dB |
| Conversion Loss Temperature Coefficient | TCCL | TC $=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 0.008 |  | $\mathrm{dB} /{ }^{\circ} \mathrm{C}$ |
| Input 1dB Compression Point | IP1dB | (Note 10) | 19 |  | dBm |
| Third-Order Input Intercept Point | IIP3 | $\begin{aligned} & \text { fiF1 }=200 \mathrm{MHz}, \text { fiF2 }=201 \mathrm{MHz}, \\ & \text { PIF1 }=\text { PIF2 }=0 \mathrm{dBm} / \text { tone } \end{aligned}$ | 29.5 |  | dBm |
| IIP3 Variation with TC |  | $\begin{aligned} & \text { fIF1 }=200 \mathrm{MHz}, \mathrm{fIF} 2=201 \mathrm{MHz}, \\ & \text { PIF1 }=\text { PIF2 }=0 \mathrm{dBm} / \text { tone, fLO }=2400 \mathrm{MHz}, \\ & \text { PLO }=0 \mathrm{dBm}, \mathrm{TC}=-40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C} \end{aligned}$ | $\pm 0.75$ |  | dB |
| LO $\pm 2 \mathrm{IF}$ Spur Rejection | $1 \times 2$ | LO-2IF | 72 |  | dBc |
|  |  | LO + 2IF | 70 |  |  |
| LO $\pm$ 3IF Spur Rejection | $1 \times 3$ | LO-3IF | 73 |  | dBc |
|  |  | LO + 3IF | 70 |  |  |
| Output Noise Floor |  | POUT $=0 \mathrm{dBm}$ ( Note 9) | -160 |  | $\mathrm{dBm} / \mathrm{Hz}$ |

Note 5: Not production tested.
Note 6: Operation outside this range is possible, but with degraded performance of some parameters. See the Typical Operating Characteristics.
Note 7: All limits reflect losses of external components, including a 0.5 dB loss at $\mathrm{f} \mid \mathrm{F}=300 \mathrm{MHz}$ due to the $1: 1$ impedance transformer. Output measurements were taken at IF outputs of the Typical Application Circuit.
Note 8: 100\% production tested for functional performance.
Note 9: Measured with external LO source noise filtered so that the noise floor is $-174 \mathrm{dBm} / \mathrm{Hz}$. This specification reflects the effects of all SNR degradations in the mixer including the LO noise, as defined in Application Note 2021: Specifications and Measurement of Local Oscillator Noise in Integrated Circuit Base Station Mixers.
Note 10: Maximum reliable continuous input power applied to the RF port of this device is +20 dBm from a $50 \Omega$ source.

# SiGe High－Linearity， 2000 MHz to 3000 MHz Upconversion／Downconversion Mixer with LO Buffer 

Typical Operating Characteristics
（Typical Application Circuit with tuning elements outlined in Table 1，Vcc＝＋5．0V，fRF＞fLO，fIF $=300 \mathrm{MHz}, \mathrm{PrF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$ ， TC $=+25^{\circ} \mathrm{C}$ ，unless otherwise noted．）

## ＋5．0V Downconverter Curves



INPUT IP3 vs．RF FREQUENCY








## SiGe High-Linearity, 2000 MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer

(Typical Application Circuit with tuning elements outlined in Table 1, Vcc $=\mathbf{+ 5 . 0 V}$, fRF $>f \mathrm{fLO}, \mathrm{fIF}=300 \mathrm{MHz}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, $\mathrm{T}_{\mathrm{C}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +5.0V Downconverter Curves





NOISE FIGURE vs. RF FREQUENCY



NOISE FIGURE vs. RF FREQUENCY





# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 1, Vcc = +5.0V, fRF >fLO, fiF $=300 \mathrm{MHz}, \mathrm{PrF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, TC $=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +5.0V Downconverter Curves



## SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 1, Vcc $=\mathbf{+ 5 . 0 V}$, fRF $>f \mathrm{fLO}, \mathrm{fIF}=300 \mathrm{MHz}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, $\mathrm{T}_{\mathrm{C}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)
+5.0V Downconverter Curves





LO PORT RETURN LOSS vs. LO FREQUENCY



SUPPLY CURRENT vs. TEMPERATURE (Tc)


# SiGe High-Linearity, 2000 MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 1, Vcc = +3.3V, fRF >fLO, fiF $=300 \mathrm{MHz}, \mathrm{PrF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, TC $=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +3.3V Downconverter Curves











## SiGe High-Linearity, 2000 MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 1, Vcc $=\mathbf{+ 3 . 3 V}$, $\mathrm{fRF}>\mathrm{fLO}, \mathrm{fIF}=300 \mathrm{MHz}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, $\mathrm{T}_{\mathrm{C}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +3.3V Downconverter Curves











# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 1, Vcc = +3.3V, fRF >fLO, fiF $=300 \mathrm{MHz}, \mathrm{PrF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, TC $=+25^{\circ} \mathrm{C}$, unless otherwise noted.)
+3.3V Downconverter Curves


RF-TO-IF ISOLATION
vs. RF FREQUENCY


LO LEAKAGE AT RF PORT vs. LO FREQUENCY


LO LEAKAGE AT IF PORT vs. LO FREQUENCY


RF-TO-IF ISOLATION
vs. RF FREQUENCY


LO LEAKAGE AT RF PORT vs. LO FREQUENCY


LO LEAKAGE AT IF PORT vs. LO FREQUENCY


RF-TO-IF ISOLATION vs. RF FREQUENCY


LO LEAKAGE AT RF PORT vs. LO FREQUENCY


## SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer

(Typical Application Circuit with tuning elements outlined in Table 1, Vcc $=\mathbf{+ 3 . 3 V}$, $\mathrm{fRF}>\mathrm{fLO}, \mathrm{fIF}=300 \mathrm{MHz}, \mathrm{PRF}=0 \mathrm{dBm}, \mathrm{PLO}=0 \mathrm{dBm}$, $\mathrm{T}_{\mathrm{C}}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)
+3.3V Downconverter Curves





LO PORT RETURN LOSS vs. LO FREQUENCY


IF PORT RETURN LOSS
vs. IF FREQUENCY


SUPPLY CURRENT
vs. TEMPERATURE


# SiGe High-Linearity, 2000 MHz to $\mathbf{3 0 0 0} \mathbf{M H z}$ Upconversion/Downconversion Mixer with LO Buffer 

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 2, $\mathbf{V c c}=+5.0 \mathrm{~V}, \mathrm{fRF}=\mathrm{fLO}+\mathrm{ffF}_{\mathrm{IF}}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ OdBm, $\mathrm{TC}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +5.0V Upconverter Curves








CONVERSION LOSS vs. RF FREQUENCY


INPUT IP3 vs. RF FREQUENCY


## SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 2, $\mathbf{V c c}=\mathbf{+ 5 . 0 V}, \mathrm{fRF}=\mathrm{fLO}+\mathrm{fIF}_{\mathrm{f}}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ OdBm, $\mathrm{T} \mathrm{C}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +5.0V Upconverter Curves











# SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer 

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 2, Vcc $=\mathbf{+ 5 . 0 V}, f \mathrm{fRF}=\mathrm{fLO}+\mathrm{ffF}_{\mathrm{IF}}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ OdBm, $\mathrm{T} \mathrm{C}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +5.0V Upconverter Curves



## SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer

(Typical Application Circuit with tuning elements outlined in Table 2, Vcc = +5.0V, fRF $=f \mathrm{fLO}+\mathrm{fIF}_{\mathrm{f}}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ OdBm, $\mathrm{TC}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +5.0V Upconverter Curves



# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 2, $\mathbf{V c c}=+3.3 \mathrm{~V}, \mathrm{fRF}=\mathrm{fLO}+\mathrm{ffF}_{\mathrm{IF}}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ OdBm, $\mathrm{T} \mathrm{C}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +3.3V Upconverter Curves








CONVERSION LOSS vs. RF FREQUENCY


INPUT IP3 vs. RF FREQUENCY


## SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 2, $\mathbf{V c c}=+3.3 \mathrm{~V}, \mathrm{fRF}=\mathrm{fLO}+\mathrm{fIF}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ OdBm, $\mathrm{T} \mathrm{C}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +3.3V Upconverter Curves




LO+2IF RESPONSE vs. RF FREQUENCY








# SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer 

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 2, Vcc $=\mathbf{+ 3 . 3 V}, \mathrm{fRF}=\mathrm{fLO}+\mathrm{fIF}_{\mathrm{f}}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ $0 \mathrm{dBm}, \mathrm{T} \mathrm{C}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +3.3V Upconverter Curves



## SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer

Typical Operating Characteristics (continued)
(Typical Application Circuit with tuning elements outlined in Table 2, $\mathbf{V c c}=+3.3 \mathrm{~V}, \mathrm{fRF}=\mathrm{fLO}+\mathrm{fIF}, \mathrm{fIF}=200 \mathrm{MHz}, \mathrm{PIF}=0 \mathrm{dBm}, \mathrm{PLO}=$ OdBm, $\mathrm{TC}=+25^{\circ} \mathrm{C}$, unless otherwise noted.)

## +3.3V Upconverter Curves






# SiGe High-Linearity, 2000 MHz to $\mathbf{3 0 0 0} \mathbf{M H z}$ Upconversion/Downconversion Mixer with LO Buffer 

Pin Description

| PIN | NAME | FUNCTION |
| :---: | :---: | :--- |
| $1,6,8,14$ | VCC | Power Supply. Bypass to GND with $0.01 \mu$ F capacitors as close as possible to the pin. |
| 2 | RF | Single-Ended $50 \Omega$ RF Input. Internally matched and DC shorted to GND through a balun. Provide <br> a DC-blocking capacitor if required. Capacitor also provides some RF match tuning. |
| $3,4,5,10$, <br> $12,13,17$ | GND | Ground. Internally connected to the exposed pad. Connect all ground pins and the exposed pad <br> (EP) together. |
| 7 | LOBIAS | LO Amplifier Bias Control. Output bias resistor for the LO buffer. Connect a $698 \Omega \pm 1 \%$ resistor (nomi- <br> nal bias condition) from LOBIAS to ground. The maximum current seen by this resistor is 3mA. |
| 9,15 | GND | Ground. Not internally connected. Ground these pins or leave unconnected. |
| 11 | LO | Local Oscillator Input. This input is internally matched to 50 $\Omega$. Requires an input DC-blocking <br> capacitor. Capacitor also provides some LO match tuning. |
| 16,20 | GND | Ground. Connect all ground pins and the exposed pad (EP) together. |
| 18,19 | IF-, IF+ | Mixer Differential IF Output/Input |
| - | EP | Exposed Pad. Internally connected to GND. Solder this exposed pad to a PCB pad that uses <br> multiple ground vias to provide heat transfer out of the device into the PCB ground planes. These <br> multiple via grounds are also required to achieve the noted RF performance. |

# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 


#### Abstract

Detailed Description When used as a low-side LO injection mixer in the 2300 MHz to 2900 MHz band, the MAX2042 provides +36 dBm of IIP3, with typical noise figure and conversion loss values of only 7.3 dB and 7.2 dB , respectively. The integrated baluns and matching circuitry allow for $50 \Omega$ single-ended interfaces to the RF and the LO ports. The integrated LO buffer provides a high drive level to the mixer core, reducing the LO drive required at the MAX2042's input to a -3 dBm to +3 dBm range. The IF port incorporates a differential interface, which is ideal for providing enhanced 2RF-2LO performance. Specifications are guaranteed over broad frequency ranges to allow for use in WCS, LTE, WiMAX, and MMDS base stations. The MAX2042 is specified to operate over an RF input range of 2000 MHz to 3000 MHz , an LO range of 1800 MHz to 2800 MHz , and an IF range of 50 MHz to 500 MHz . The external IF transformer sets the lower frequency range (see the Typical Operating Characteristics for details). Operation beyond these ranges is possible (see the Typical Operating Characteristics for additional information).


RF Interface and Balun
The MAX2042 RF input provides a $50 \Omega$ match when combined with a series DC-blocking capacitor. This DC-blocking capacitor required as the input is internally DC shorted to ground through the on-chip balun. When using an 8.2 pF DC-blocking capacitor, the RF port input return loss is typically 15 dB over the RF frequency range of 2500 MHz to 2900 MHz .

## LO Inputs, Buffer, and Balun

The MAX2042 is optimized for low-side LO injection applications with an 1800 MHz to 2800 MHz LO frequency range. The LO input is internally matched to $50 \Omega$, requiring only a 2 pF DC-blocking capacitor. A two-stage internal LO buffer allows for a -3 dBm to +3 dBm LO input power range. The on-chip low-loss balun, along with an LO buffer, drives the double-balanced mixer. All interfacing and matching components from the LO inputs to the IF outputs are integrated on-chip.

## High-Linearity Mixer

The core of the MAX2042 is a double-balanced, highperformance passive mixer. Exceptional linearity is provided by the large LO swing from the on-chip LO buffer. IIP3, 2RF-2LO rejection, and noise-figure performance are typically $+36 \mathrm{dBm}, 70 \mathrm{dBc}$, and 7.3 dB , respectively.

Differential IF Interface
The MAX2042 has an IF frequency range of 50 MHz to 500 MHz , where the low-end frequency depends on the frequency response of the external IF components.
The MAX2042's differential ports are ideal for providing enhanced 2RF-2LO performance. The user can use a differential IF amplifier or SAW filter on the mixer IF port, but a DC block is required on both IF+/IF- ports to keep external DC from entering the IF ports of the mixer. Typical applications typically use a 1:1 transformer such as the MABAES0029 to transform the $50 \Omega$ differential interface to a $50 \Omega$ single-ended interface. The loss of this transformer is included in the data presented in this data sheet. In addition, the IF interface directly supports single-ended AC-coupled signals into or out of IF+ by shorting IF- to ground, and a $1 \mathrm{k} \Omega$ resistor from IF+ to ground.

## Applications Information

## Input and Output Matching

The RF input provides a $50 \Omega$ match when combined with a series DC-blocking capacitor. Use an 8.2 pF capacitor value for RF frequencies ranging from 2000 MHz to 3000 MHz . The LO input is internally matched to $50 \Omega$; use a 2 pF DC-blocking capacitor to cover operations spanning the 1800 MHz to 2800 MHz range. The IF output impedance is $50 \Omega$ (differential). For evaluation, an external low-loss 1:1 (impedance ratio) balun transforms this impedance down to a $50 \Omega$ single-ended output (see the Typical Application Circuit).

Reduced-Power Mode
The MAX2042 has one pin (LOBIAS) that allows an external resistor to set the internal bias current. A nominal value for this resistor is shown in Tables 1 and 2. Larger value resistors can be used to reduce power dissipation at the expense of some performance loss. See the Typical Operating Characteristics to evaluate the power vs. performance tradeoff. If $\pm 1 \%$ resistors are not readily available, substitute with $\pm 5 \%$ resistors.
Significant reductions in power consumption can also be realized by operating the mixer with an optional supply voltage of +3.3 V . Doing so reduces the overall power consumption by up to $43 \%$. See the +3.3 V Supply AC Electrical Characteristics table and the relevant +3.3 V curves in the Typical Operating Characteristics section to evaluate the power vs. performance tradeoffs.

# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

## Layout Considerations

A properly designed PCB is an essential part of any RF/microwave circuit. Keep RF signal lines as short as possible to reduce losses, radiation, and inductance. For the best performance, route the ground pin traces directly to the exposed pad under the package. The PCB exposed pad MUST be connected to the ground plane of the PCB. It is suggested that multiple vias be used to connect this pad to the lower-level ground planes. This method provides a good RF/thermal conduction path for the device. Solder the exposed pad on the bottom of the device package to the PCB. The MAX2042 evaluation kit can be used as a reference for board layout. Gerber files are available upon request at www.maxim-ic.com.

Power-Supply Bypassing
Proper voltage-supply bypassing is essential for highfrequency circuit stability. Bypass each VCC pin with the capacitors shown in the Typical Application Circuit and see Tables 1 and 2.

Exposed Pad RF/Thermal Considerations The exposed pad (EP) of the MAX2042's 20-pin thin QFN package provides a low thermal-resistance path to the die. It is important that the PCB on which the MAX2042 is mounted be designed to conduct heat from the EP. In addition, provide the EP with a low-inductance path to electrical ground. The EP MUST be soldered to a ground plane on the PCB, either directly or through an array of plated via holes.

## Table 1. Downconverter Mode Component Values

| DESIGNATION | QTY | DESCRIPTION | COMPONENT SUPPLIER |
| :---: | :---: | :--- | :--- |
| C1 | 1 | 8.2 pF microwave capacitor (0402) | Murata Electronics North America, Inc. |
| C2, C6, C8, C11 | 4 | $0.01 \mu F$ microwave capacitors (0402) | Murata Electronics North America, Inc. |
| C3, C9 | 0 | Not installed, capacitors | - |
| C5 | 0 | Not installed, capacitor | - |
| C10 | 1 | $2 p F$ microwave capacitor (0402) | Murata Electronics North America, Inc. |
| R1 | 1 | $698 \Omega \pm 1 \%$ resistor (0402) | Digi-Key Corp. |
| T1 | 1 | $1: 1$ IF balun MABAES0029 | M/A-Com, Inc. |
| U1 | 1 | MAX2042 IC (20 TQFN) | Maxim Integrated Products, Inc. |

Table 2. Upconverter Mode Component Values

| DESIGNATION | QTY | DESCRIPTION | COMPONENT SUPPLIER |
| :---: | :---: | :--- | :--- |
| C1 | 1 | 8.2 pF microwave capacitor (0402) | Murata Electronics North America, Inc. |
| C2, C6, C8, C11 | 4 | $0.01 \mu F$ microwave capacitors (0402) | Murata Electronics North America, Inc. |
| C3, C9 | 0 | Not installed, capacitors | - |
| C5 | 0 | Not installed, capacitor | - |
| C10 | 1 | $2 p F$ microwave capacitor (0402) | Murata Electronics North America, Inc. |
| R1 | 1 | $698 \Omega \pm 1 \%$ resistor (0402) | Digi-Key Corp. |
| T1 | 1 | $1: 1$ IF balun MABAES0029 | M/A-Com, Inc. |
| U1 | 1 | MAX2042 IC (20 TQFN) | Maxim Integrated Products, Inc. |

# SiGe High-Linearity, 2000 MHz to 3000 MHz Upconversion/Downconversion Mixer with LO Buffer 



# SiGe High-Linearity, 2000MHz to 3000MHz Upconversion/Downconversion Mixer with LO Buffer 

Chip Information
PROCESS: SiGe BiCMOS

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages

| PACKAGE TYPE | PACKAGE CODE | DOCUMENT NO. |
| :---: | :---: | :---: |
| 20 TQFN-EP | $\mathrm{T} 2055+3$ | $\underline{\underline{21-0140}}$ |

